

## Primary-Side Control for DCM Flyback Power Supply

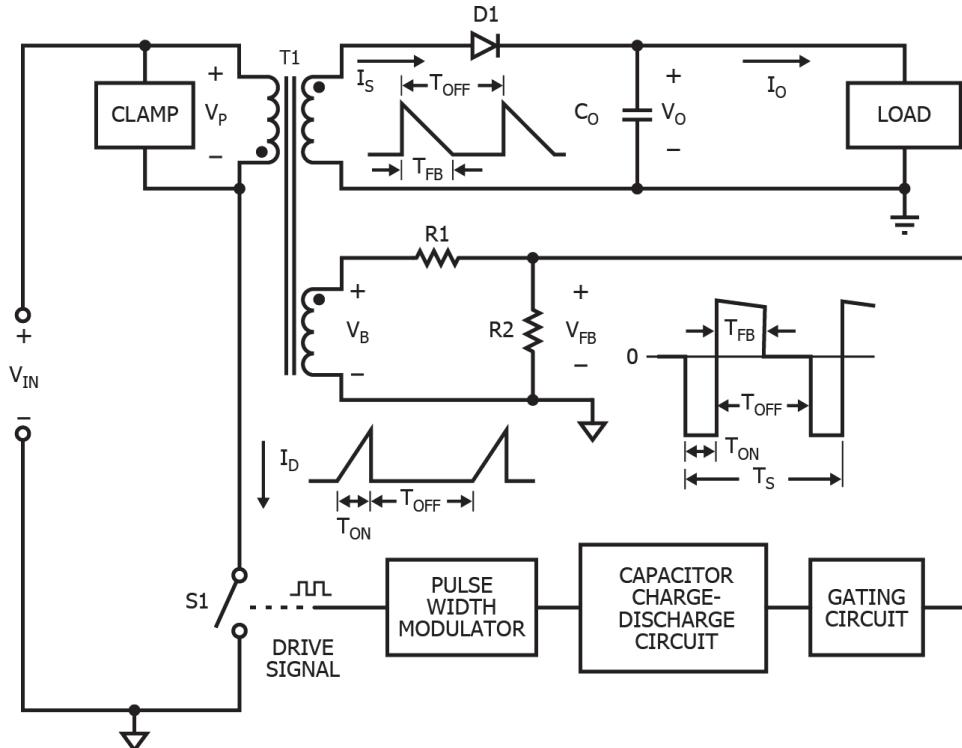
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### Summary of the Idea

A primary-side controller for an isolated flyback power supply operating in discontinuous conduction mode (DCM) may use a primary-side winding to sense and to regulate the output voltage as shown in Figure 1. The sense winding provides a switching feedback voltage  $V_{FB}$  that is gated to a capacitor charge-discharge circuit during a time  $T_{FB}$  when the output rectifier is conducting. A capacitor discharges when  $V_{FB}$  is greater than a desired regulation value  $V_{REF}$ . The capacitor charges when the  $V_{FB}$  is less than  $V_{REF}$ . The voltage on the capacitor at the end of  $T_{FB}$  determines the duty ratio of a pulse width modulator (PWM) that drives the primary switch. For stable operation and fast transient response, the capacitor must discharge to a constant reset value  $V_{RESET}$  when  $V_{FB}$  is greater than  $V_{REF}$ .

### Description

Figure 1 shows that feedback voltage  $V_{FB}$  is negative when switch S1 conducts current  $I_D$  during time  $T_{ON}$



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Figure 1. An example of an isolated flyback power supply operating in discontinuous conduction mode that uses a controller with a capacitor charge-discharge circuit to process a feedback signal from a winding referenced to the primary side. A pulse width modulator drives primary switch S1 with a duty ratio that corresponds to a signal received from the capacitor charge-discharge circuit.

within a switching period  $T_S$ .  $V_{FB}$  is positive and represents the output voltage  $V_O$  during time  $T_{FB}$  when output rectifier D1 conducts secondary current  $I_S$ .

$V_{FB}$  typically has a decreasing value during time  $T_{FB}$  as secondary current  $I_S$  decreases from an initial peak value. This AC ripple voltage on DC output voltage  $V_O$  is typically dominated by current  $I_S$  with the equivalent series resistance (ESR) of capacitor  $C_0$ .

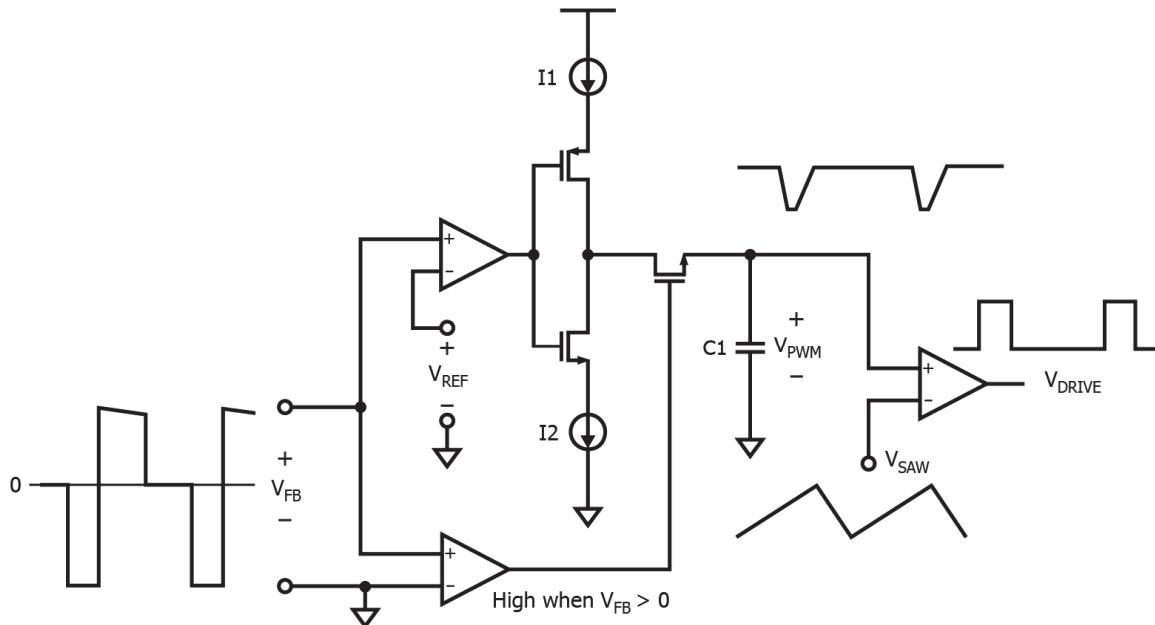
Figure 2 is an example gating, charge-discharge, and PWM circuit. Switched current sources I1 and I2 respectively charge and discharge capacitor C1 when the feedback voltage  $V_{FB}$  is less than or greater than  $V_{REF}$ , respectively. The voltage on C1 is compared to a ramping voltage  $V_{SAW}$  to produce a pulse width modulated output voltage  $V_{DRIVE}$  for primary switch S1.

Figure 3 shows the waveforms of Figure 2 in greater detail with emphasis on the relevant timing relationships.

# Circuit Idea

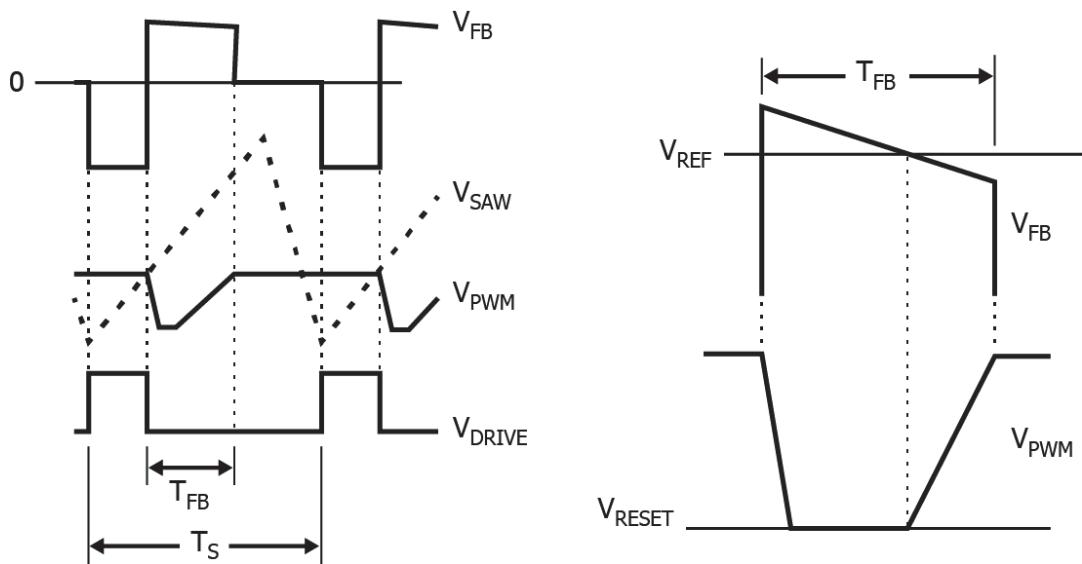
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Figure 2. An example of a capacitor charge-discharge circuit and pulse width modulator for a flyback power supply. The positive portion of feedback voltage  $V_{FB}$  is compared to a reference voltage  $V_{REF}$  to gate current sources  $I1$  and  $I2$  that charge and discharge capacitor  $C1$ . Voltage  $V_{PWM}$  on capacitor  $C1$  is compared to a ramping voltage  $V_{SAW}$  to produce a pulse width modulated drive voltage  $V_{DRIVE}$ .



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Figure 3. Waveforms showing the timing relationships of signals from the example control circuit of Figure 2. Decreasing feedback voltage  $V_{FB}$  during the feedback interval  $T_{FB}$  crosses a reference value  $V_{REF}$  that is representative of a desired value of output voltage  $V_o$ . Current source  $I2$  in the circuit of Figure 2 is chosen to discharge capacitor  $C1$  to a constant reset value  $V_{RESET}$  before current source  $I2$  charges  $C1$ . In some examples, the value of  $V_{RESET}$  may be zero volts.